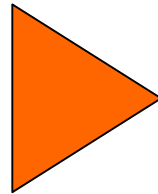
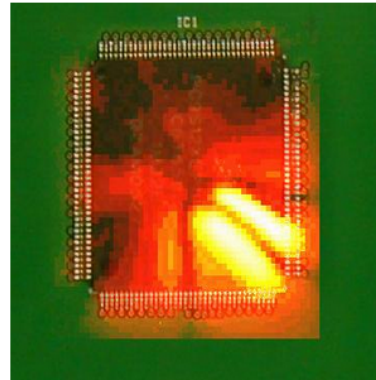


EMC of ICs Practical Trainings

www.ic-emc.org



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Welcome to IC-EMC homepage

The IC-EMC software is a non-commercial tool dealing with electromagnetic compatibility (EMC) of integrated circuits (IC), and covering both parasitic emission and susceptibility to radio-frequency interference ([more](#)).

Based on 20 years of research, IC-EMC gathers a unique collection of IC models, tools and EMC measurements related to more than 15 IC case studies compiled in a user's manual ([more](#)).

A user's manual describes the basic and advanced features of IC-EMC. Application notes described new case studies, specific tools and approaches ([more](#)). The tool is also used for trainings in university and in industry ([more](#)).

A five day training about EMC of ICs is proposed. Real case studies and practical trainings are proposed based on IC-EMC. More information ([here](#)).

A book ([more](#)) is also proposed which basic notions for learning how to model circuits and their surrounding environment (PCB) with respect to emission, immunity and signal integrity issues. The book also provides a series of exercises. Corrections can be found here. ([more](#))

<http://www.winspice.com/>

Unzip

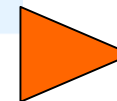
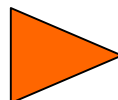
■ Unzip both files in \documents

 IC-EMC-2v9.zip

 WinSpice1v5.zip

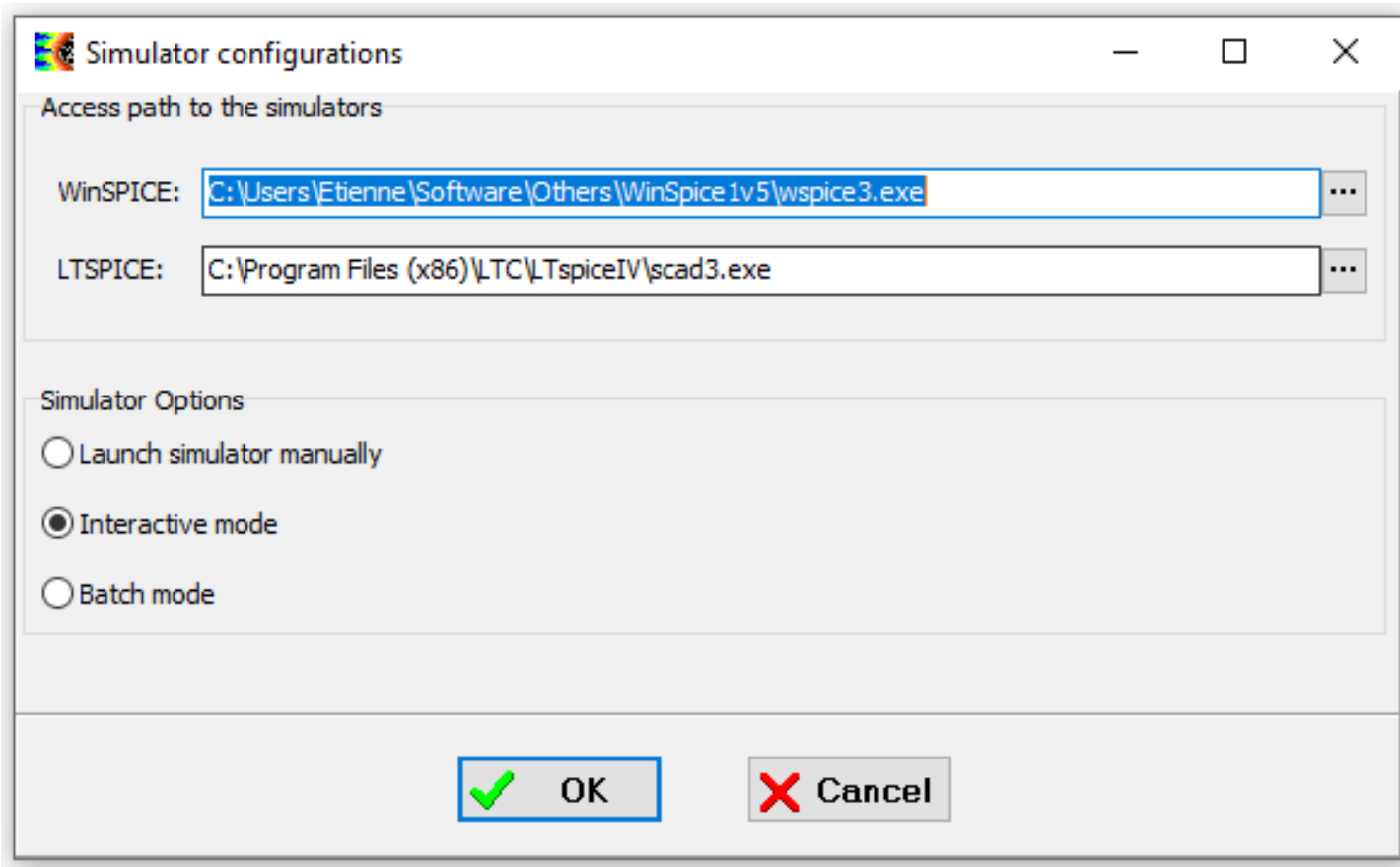
basic
book
case_study
EMC_lib
emission
eseca
examples
getting_started
html
ibis
ieee
immunity
lib
misc
msc-eseca
near_field
package
passive
st
example.cir
example.txt
ic_emc.exe

lib
DelsL1.isu
DEMO
DEMO.CIR
DEMO1.CIR
DEMO2.CIR
help.exe
LF156
license.txt
makeidx.exe
multidec.exe
proc2mod.exe
README.TXT
reg_license.txt
REGKEY.DAT
sconvert.exe
sp0.tmp
spicepp.exe
test.cir
WS_FTP.LOG
wspice3.exe



Configure IC-EMC

■ Find wspice.exe



Objectives

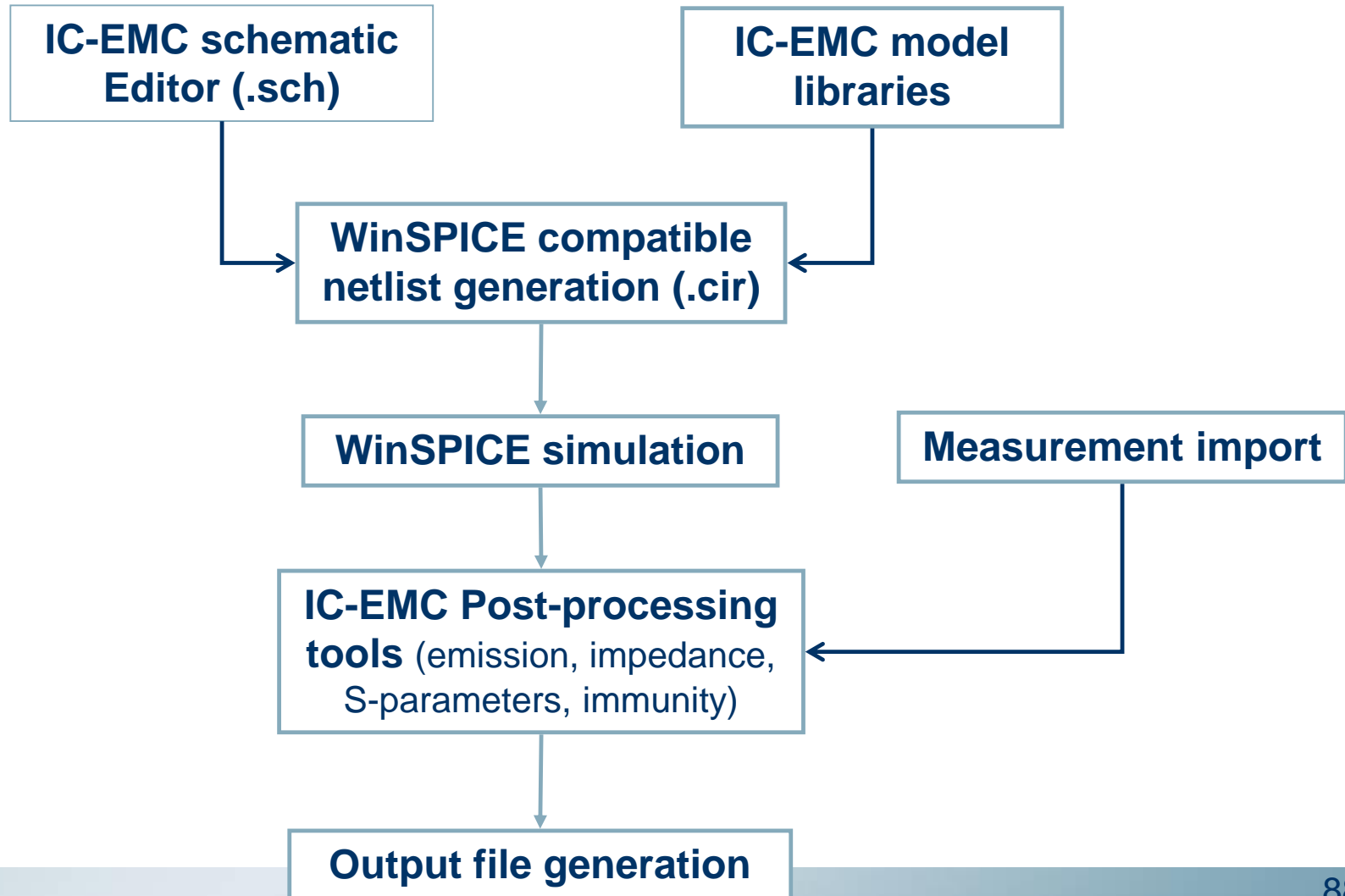
- Get familiar with IC-EMC/Winspice
- Illustrate parasitic emission mechanisms
- Understand parasitic emission reduction strategies
- Power Decoupling Network modelling
- Basis of conducted and radiated emission modelling
- Basis of immunity modelling

Summary























- **IC-EMC – Reference**
- **Ex. 1. FFT of typical signals**
- **Ex. 2. Transient current estimation**
- **Ex. 3. Interconnect parasitics**
- **Ex. 4. Impedance mismatch**
- **Ex. 5. di/dt noise**
- **Ex. 6. intrinsic decoupling**
- **Ex. 7. added on-chip decoupling**
- **Ex. 8. PDN modelling**
- **Ex. 9. Radiated emission modelling**

- **Ex. 10. Estimation of susceptibility level**
- **Ex. 11. Susceptibility of analog input**
- **Ex. 12. Susceptibility of output buffer**
- **Ex. 13. Susceptibility of a micro-controller**

IC-EMC - Simulation flow



IC-EMC – Most Important Icons

	Open schematic (.sch)		Build SPICE netlist (.cir)
	Save schematic (.sch)		Spectrum analysis
	Delete symbols		Near field emission simu.
	Copy symbols		Immunity simulation
	Move symbols		Time domain analysis
	Rotate symbols		Impedance simulation
	Flip symbols		S parameter simulation
	Add Text line		Ibis file editor
	Add a line		Parametric analysis
	View electrical net		Symbol palette
	Zoom in/out		View all schematic

IC-EMC – Link to WinSpice

- Click on WinSPICE.exe
- Click File/Open to open a circuit netlist (.cir) generated by ic-emc.

- IC-EMC main commands (text line):

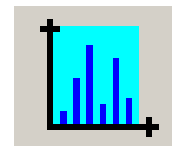
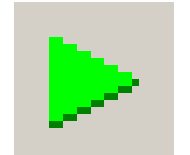
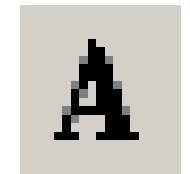
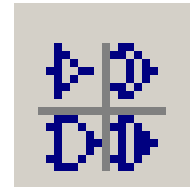


```
WinSpice v1.05.01
File Edit Settings Help
*****
WinSpice © Copyright 1996-2003 OuseTech Ltd. All Rights Reserved.
*****
Version: 1.05.01
Built : Dec 10 2003 00:47:53
Shareware version of WinSpice. For non-commercial use only.
Please read the file 'license.txt' for conditions of use.
*****
Type "help" for more information, "quit" to leave.
WinSpice 1 -> _
```

Simulation command	Command line	Parameters
Transient simulation	.tran 0.1n 100n	step + stop time
DC simulation	.DC Vdd 0 5 0.1	source + start + stop + step
Small signal freq. analysis	.AC DEC 100 1MEG 1G	sampling + nb points + start + stop
Load SPICE library	.lib 65nm.lib	Path and file name

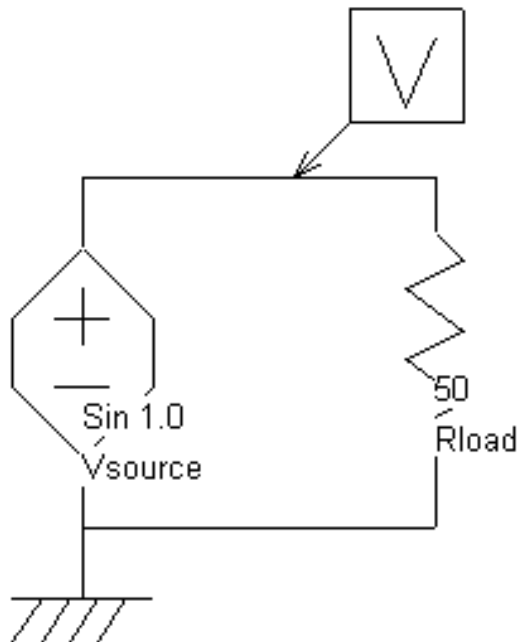
Exercise 1. FFT of typical signals

- Create the schematic
- Set the source generator
- Transient simulation
- FFT by IC-EMC
- Simulate the FFT of a sinus and a square signal

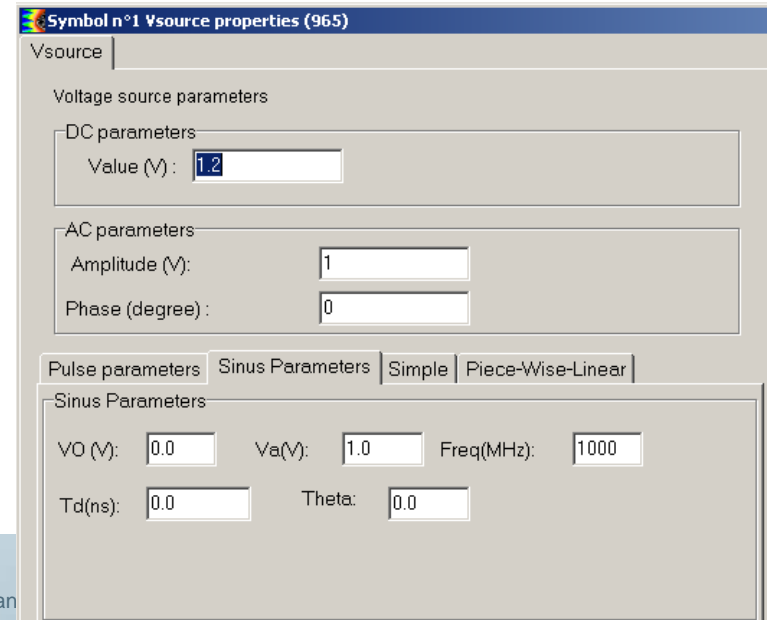


Exercise 1. FFT of typical signals

- FFT of a sinus source
 - Set the voltage generator properties:
 - Frequency = 1 GHz
 - Amplitude = 1 V



Ex1-FFT-sinus.sch



Exercise 1. FFT of typical signals

- FFT of a sinus source

- Type the simulation command: `.tran 1n 50n`



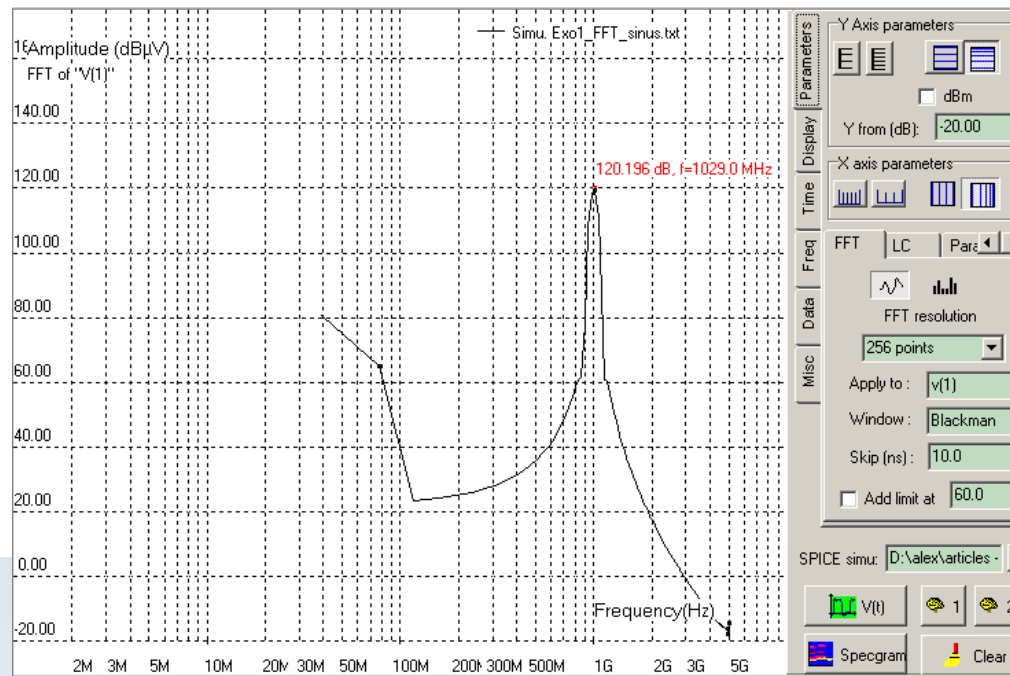
- Simulate the response in time domain.



- Compute the FFT.

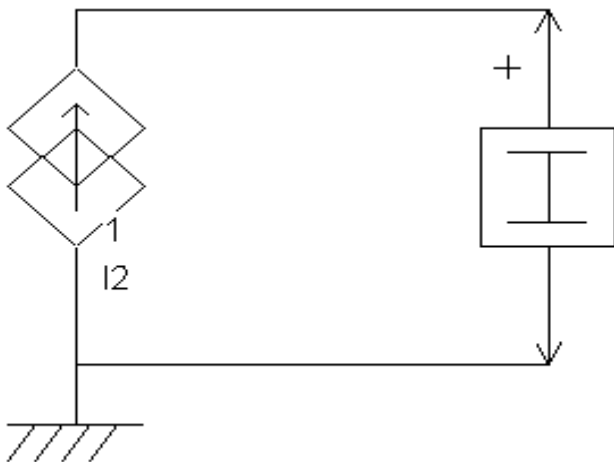


- Does the FFT result correlate with theoretical result ?

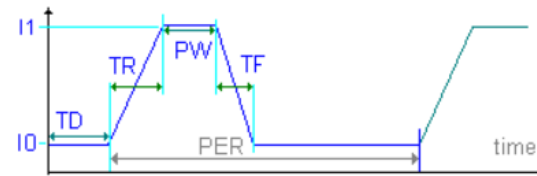
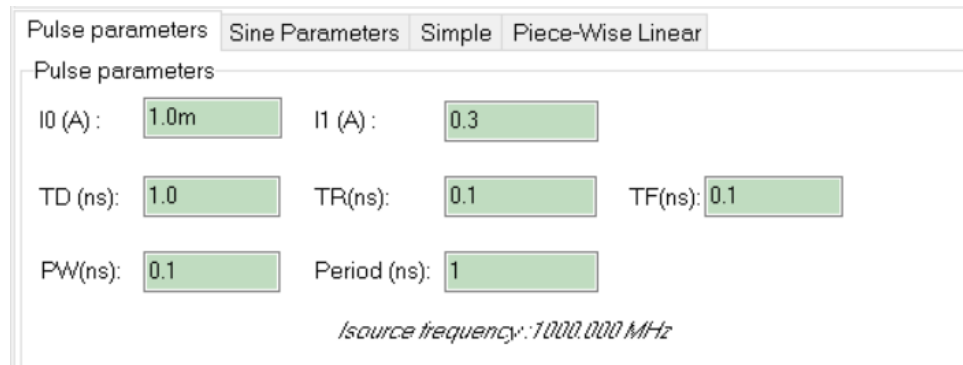


Exercise 1. FFT of typical signals

- FFT of a square current source
 - Set the generator properties
 - For example:
 - Period = 1 n,
 - PW as small as possible
 - $T_r = 0.1\text{n}$, $T_f = 0.1\text{ n}$
 - $V_0 = 0\text{ V}$, $V_1 = 0.9\text{ V}$



.tran 0.1n 500n



Ex1-FFT-Pulse.sch

Exercise 1. FFT of typical signals

- FFT of a square source

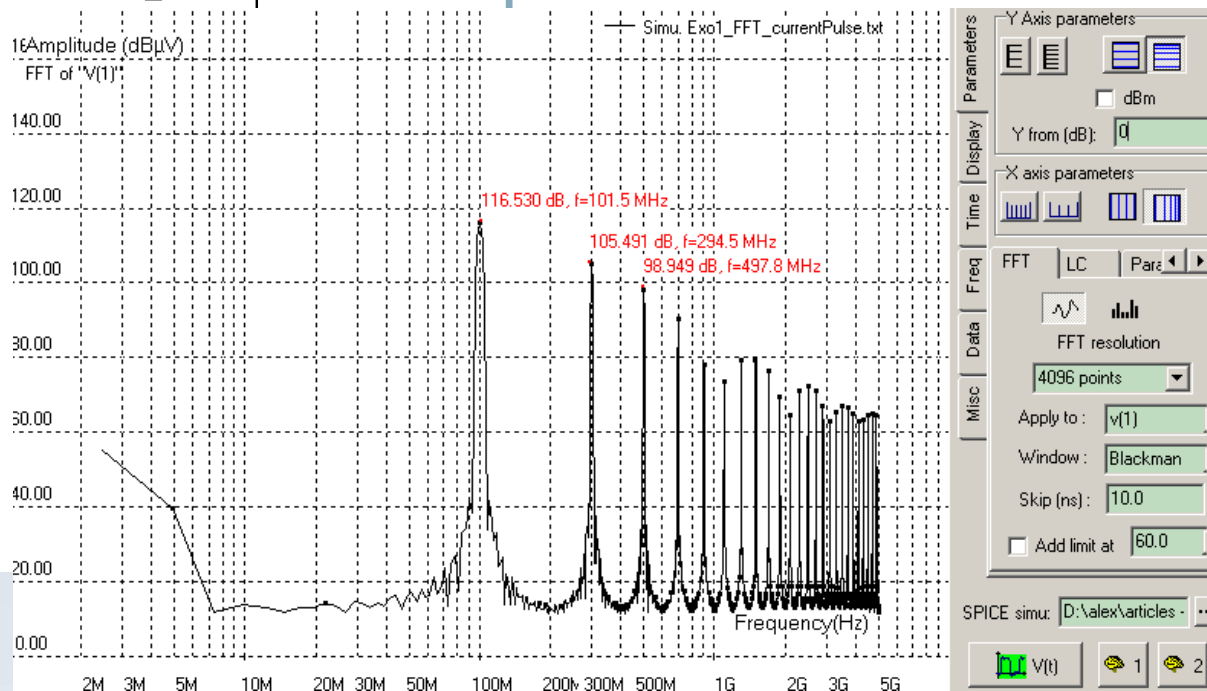
For a square signal ($T_r=0$)

$$|c_n^+| = \frac{2A\tau}{T} \left| \frac{\sin\left(n\pi \frac{\tau}{T}\right)}{n\pi \frac{\tau}{T}} \right|, n > 0$$

$$c_0 = \frac{A\tau}{T}$$

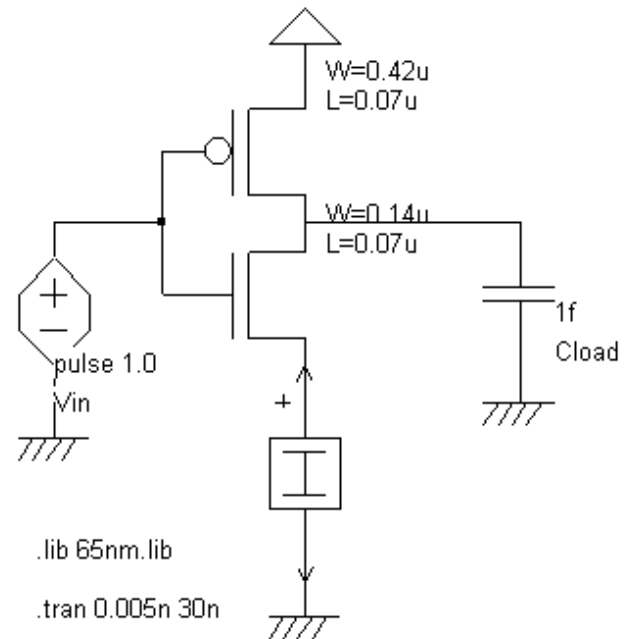
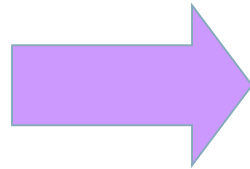
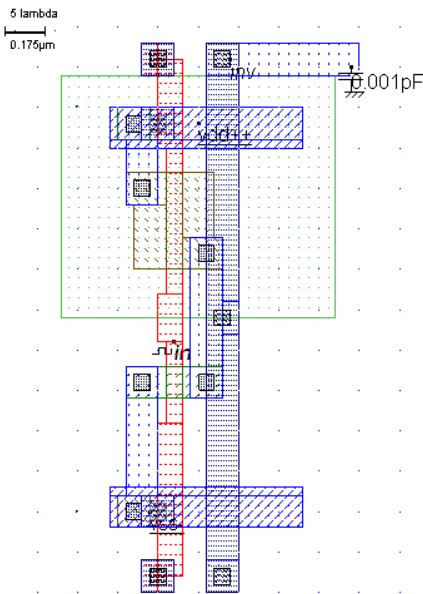
For a trapezoidal signal ($T_r=T_f$)

$$|c_n^+| = \frac{2A\tau}{T} \left| \frac{\sin\left(n\pi \frac{\tau}{T}\right)}{n\pi \frac{\tau}{T}} \right| \left| \frac{\sin\left(n\pi \frac{t_r}{T}\right)}{n\pi \frac{t_r}{T}} \right|, n > 0$$



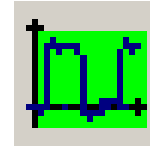
Exercise 2. Transient current estimation

- Standard cell inverter in CMOS technology
- Typical load capacitance
- Observe in time domain the current through Vss.

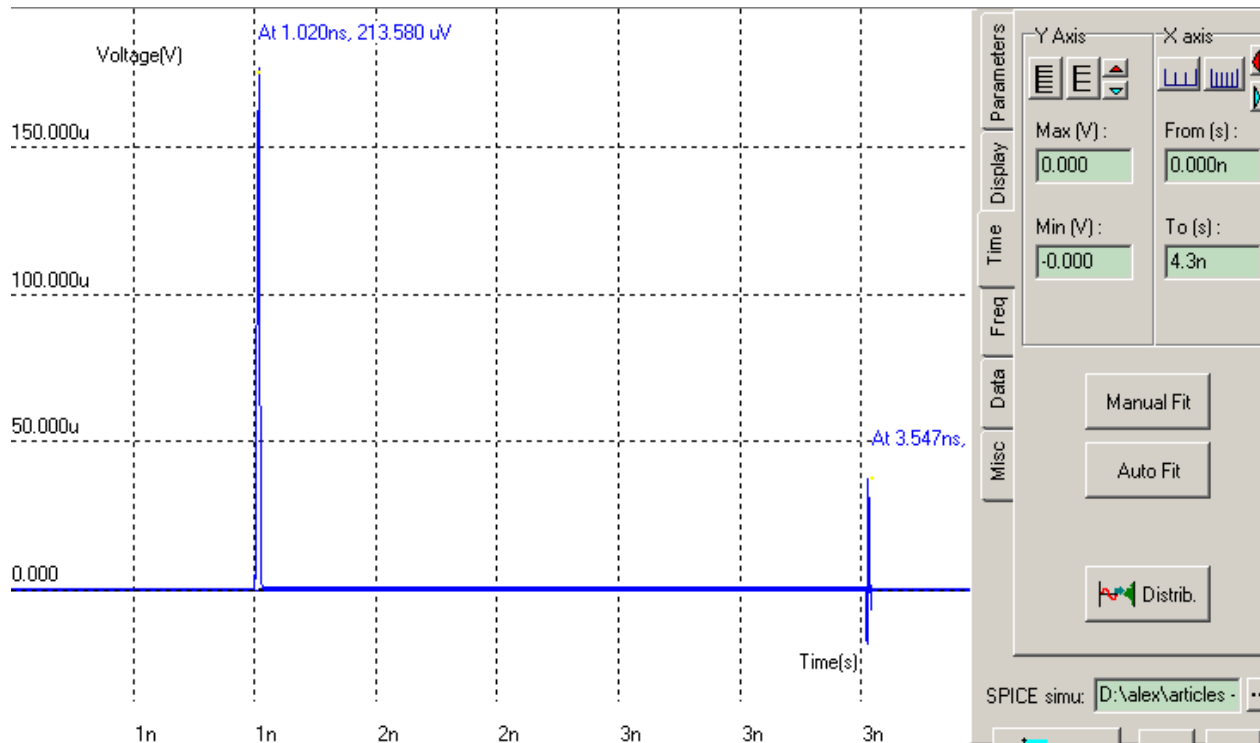
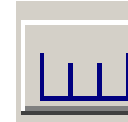


Ex2-transient_inverter.sch

Exercise 2. Transient current estimation

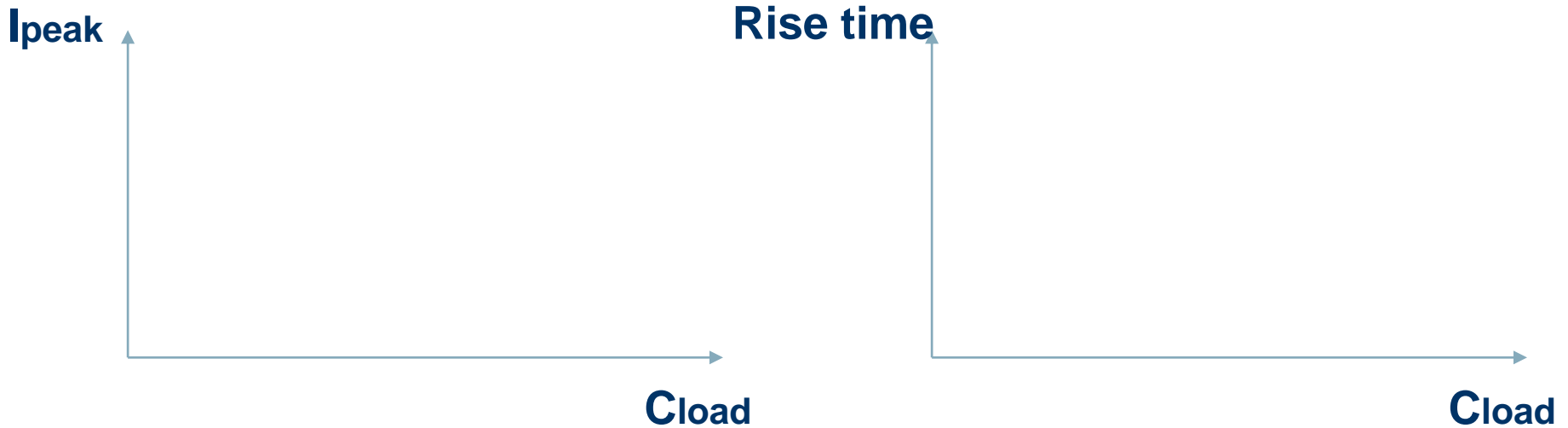


- Time domain simulation
- Adjust scales (Autofit and zoom on time axis)



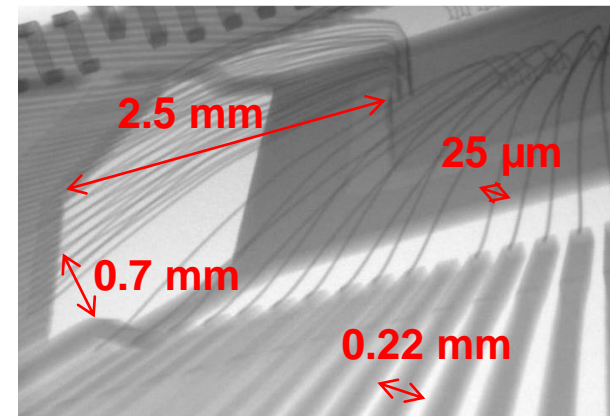
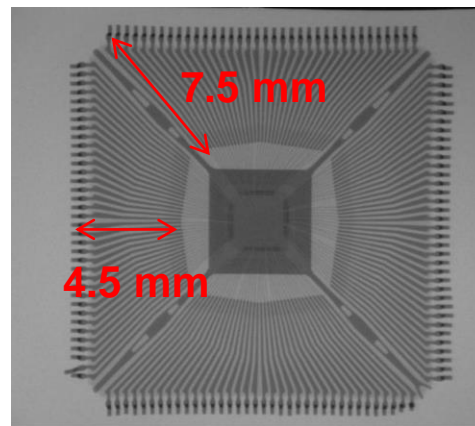
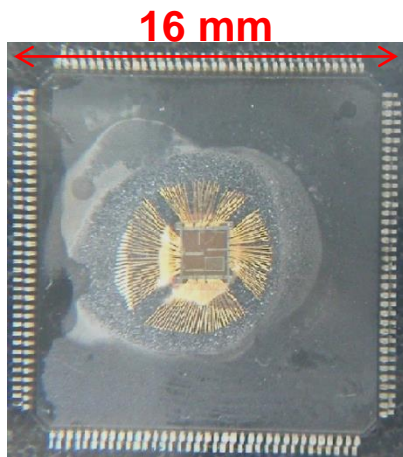
Exercise 2. Transient current estimation

- What is the influence of the load capacitance (1 fF to 1 pF) ?



Exercise 3. Interconnect parasitics

- The core is mounted in a QFP100 package.
- A pair of pins is dedicated to supply the core



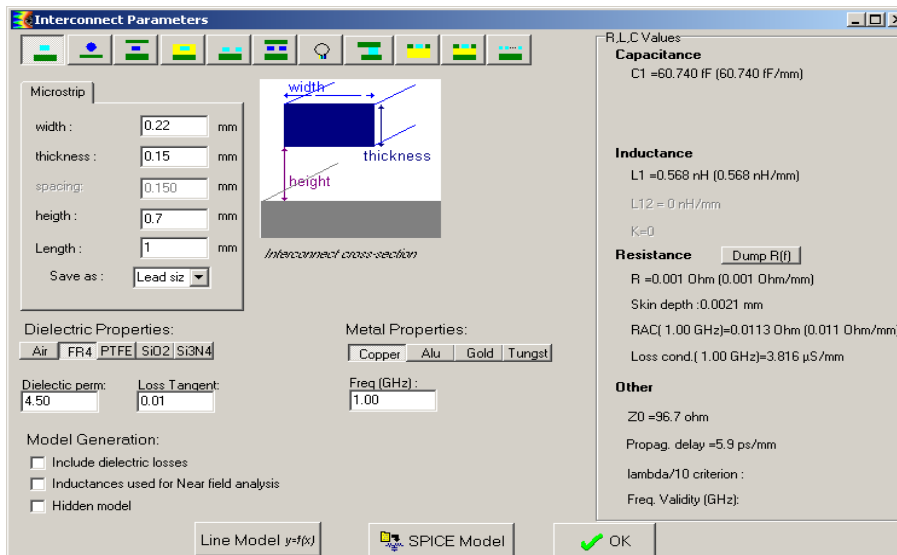
Evaluate the electrical parasitic associated to the power supply pair.

Exercise 3. Interconnect parasitics

Use Tools/Interconnects Parameters to evaluate R, L, C associated to package pins.

Empirical estimation :

- Lead : $L = 0.5$ nH/mm and $C = 0.1$ pF/mm
- Bonding : $L = 1$ nH/mm

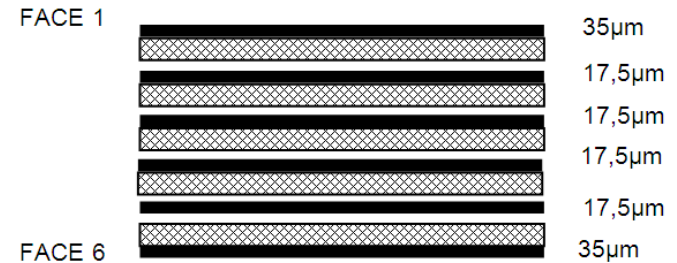


$$L = \frac{\mu_o l}{2\pi} \ln \left(\frac{8h}{W} + \frac{W}{4h} \right)$$

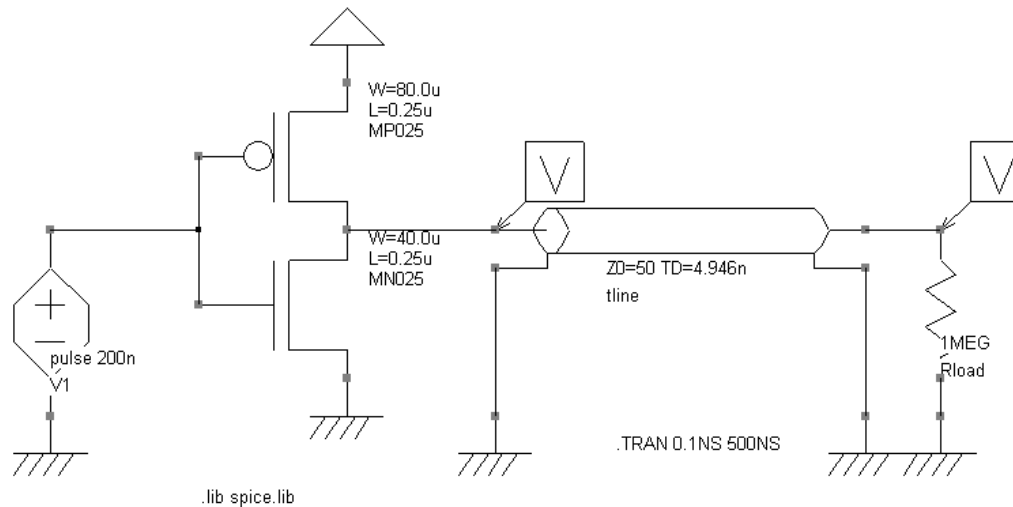
$$L = \frac{\mu_o l}{2\pi} \times \ln \left(\frac{4h}{r} \right)$$

Exercise 4. Impedance Mismatch

- Generate a fast clock
- Load a transmission line model
- Terminate by a high impedance
- Terminate by 50Ω



1.6mm total

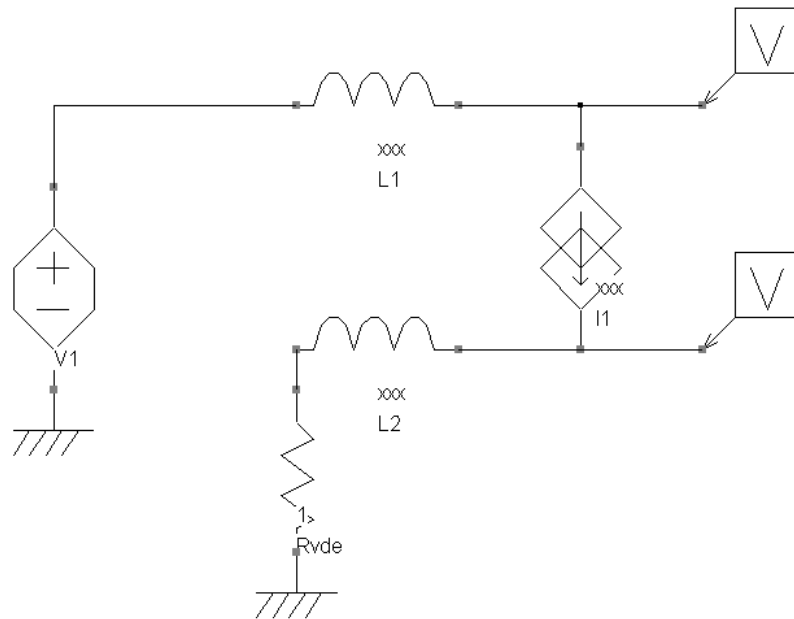


Exercise 5. di/dt noise

- Estimate the voltage bounce on Vdd and Vss pins of the core when it is mounted in a QFP 64.
- The core clock is 20 MHz.

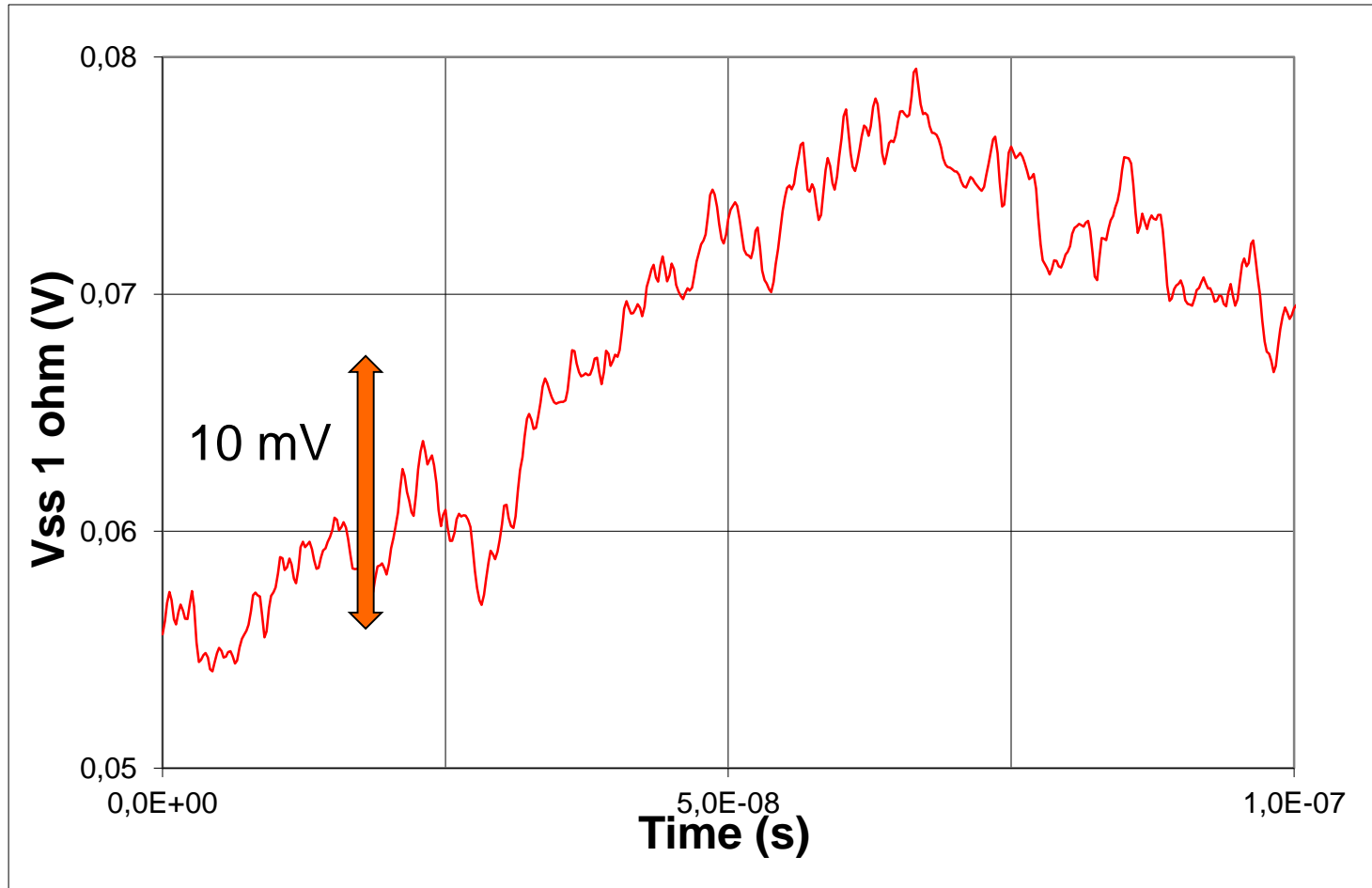
$$\Delta V = ?$$

Core noise margin ?



Ex4-didt_noise.sch

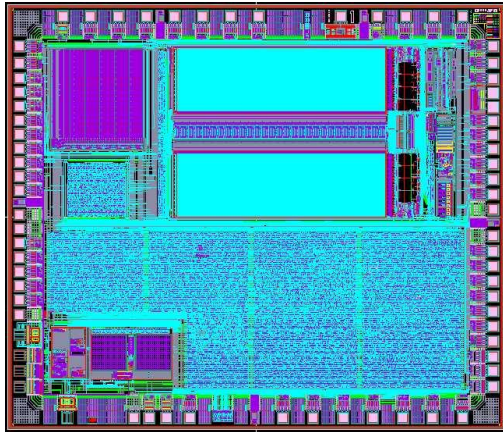
Exercise 5. di/dt noise



Exercise 6. Intrinsic decoupling

IC-EMC reference manual p. 18

Consider a synchronous digital core in CMOS 65 nm formed by 100 000 gates with the following parameters



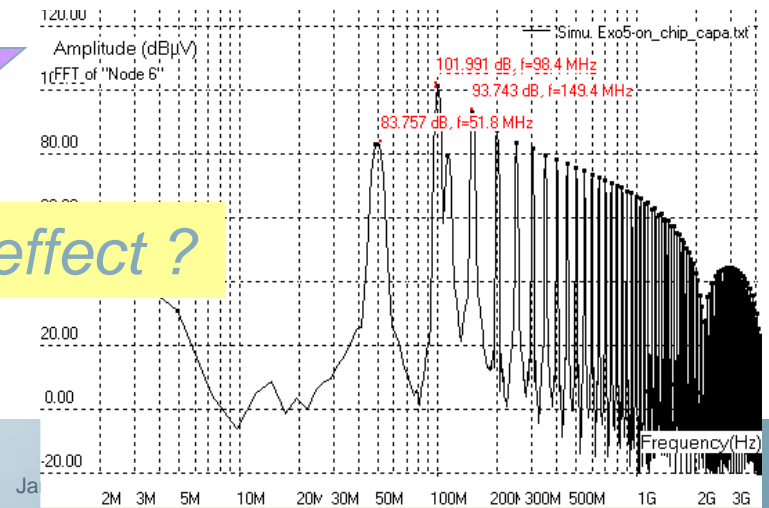
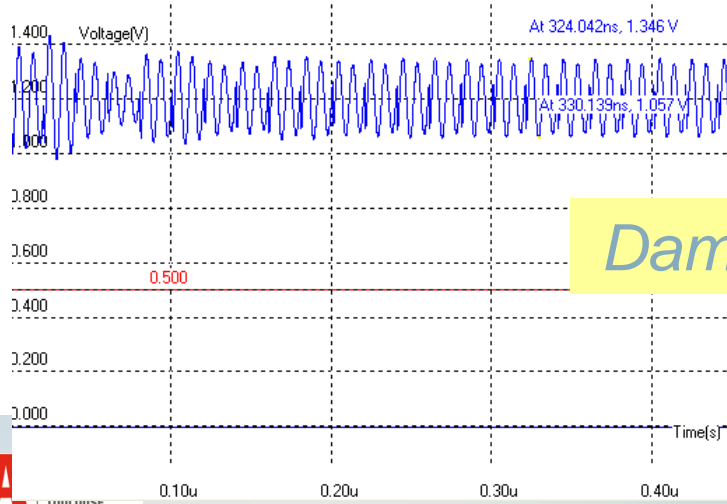
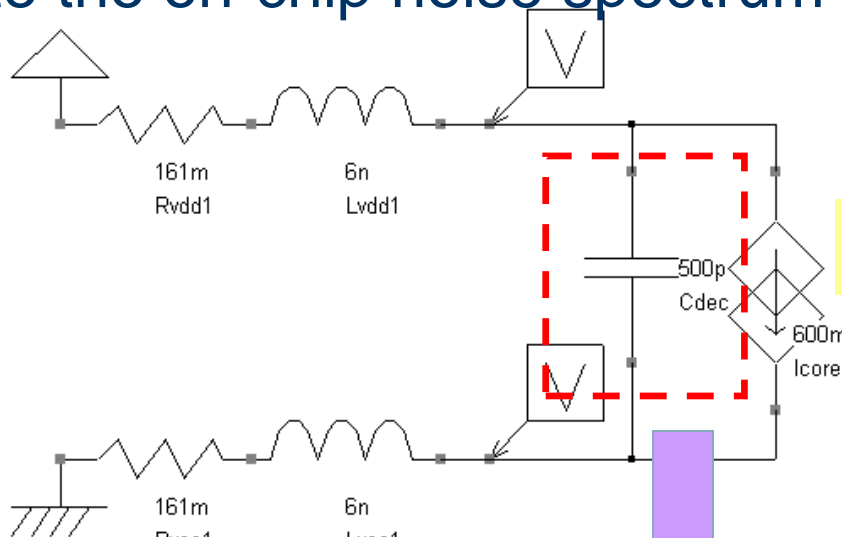
Std cell	Number	Typical input capa (fF)	Peak current / gate (μA)
Inverter	35000	1	120
NAND2	25000	1	150
DREG	20000	2	200
NOR2	20000	1	150

Estimate the intrinsic decoupling

Estimate the dynamic current consumed by the circuit.

Exercise 6. Intrinsic decoupling

Compute the on-chip noise spectrum



Damping effect ?

Exercise 7. Added on-chip decoupling

IC power supply rails parasitics

- Consider a 1 mm long and 40 μm wide Vdd or Vss line.

Metal level	Thickness	Height to substrate
M6	0.4 μm	4.5 μm
M5	0.4 μm	3.3 μm
M1	0.3 μm	0.5 μm

- Total Inductance and resistance of chip power supply rails ?

Exercise 7. Added on-chip decoupling

On-chip capacitor budget :

$$\Delta i = C \frac{\Delta v}{\Delta t} \Leftrightarrow C = \frac{\Delta i \times \Delta t}{\Delta v}$$

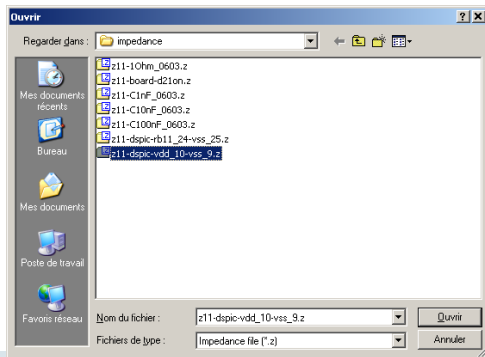
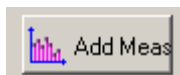
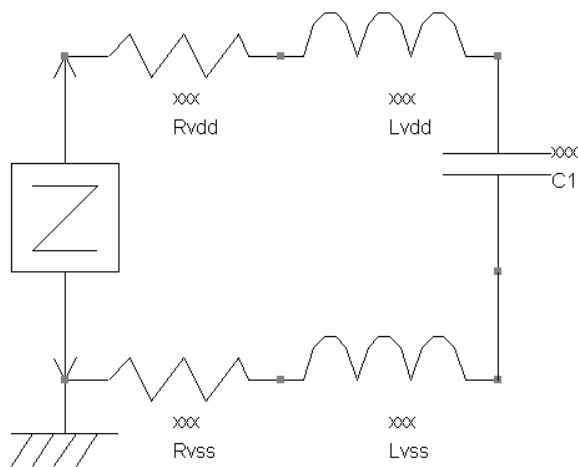
- Is the intrinsic capacitance is sufficient to reach the noise margin target ?
- How much capacitance should be added in the circuit ?

Capa. type	Capacitor density (fF/μm ²)
Capa cell	2.2
Poly1 – Poly2	1.7
MIM	1.4

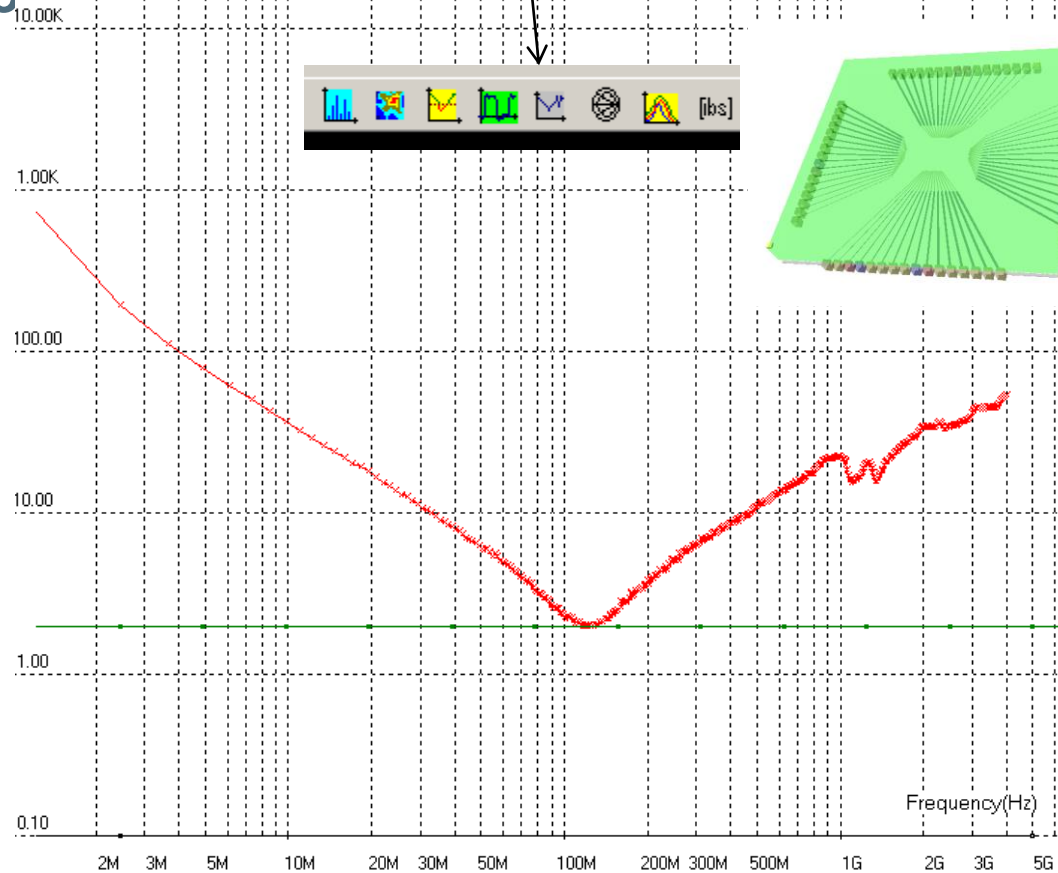
Ex6-AddRonChipC.sch

Exercise 8. PDN Modelling

- DSPIC Z(f): find an R,L,C model
- Tune to measurement file:



Impedance vs. Freq

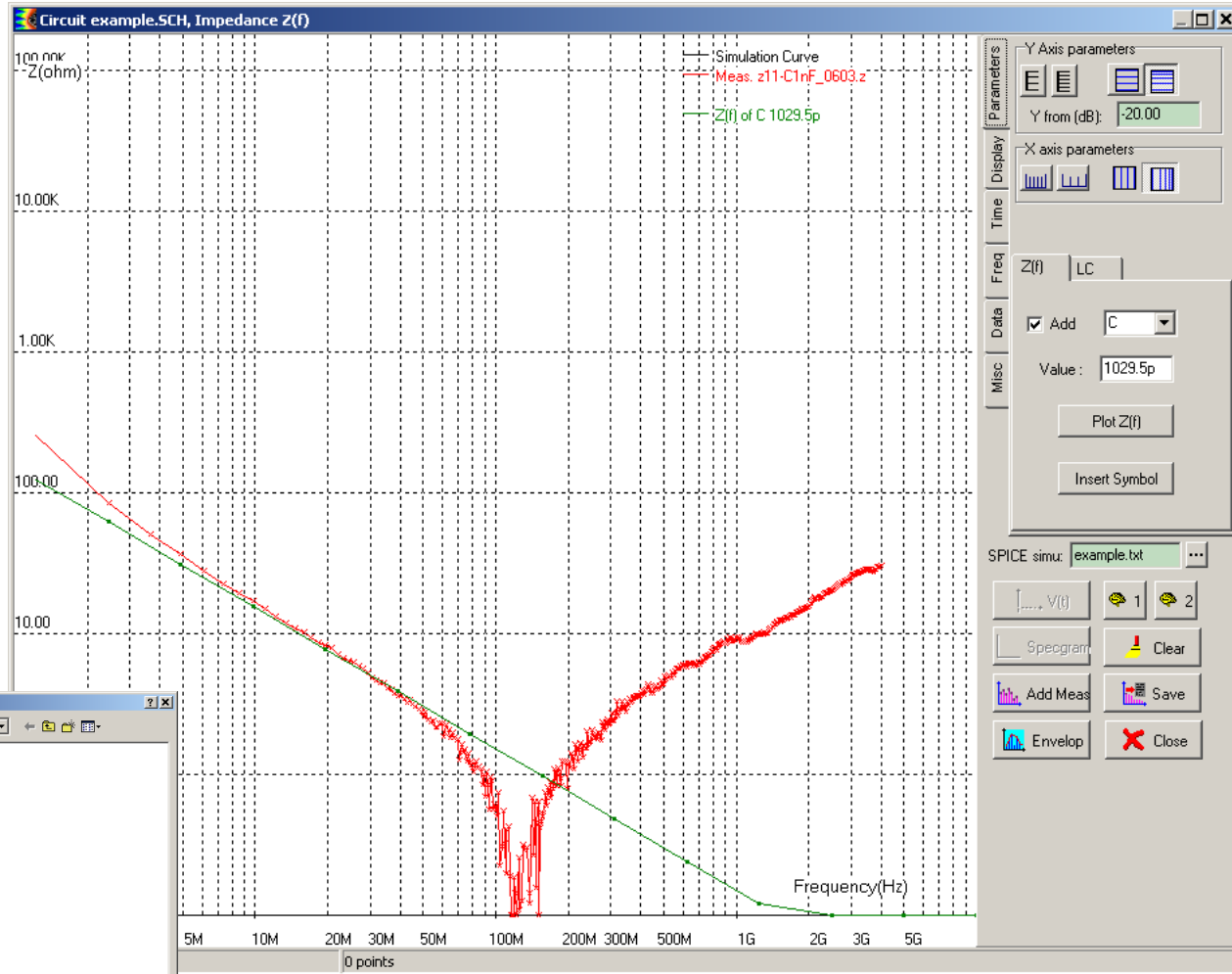
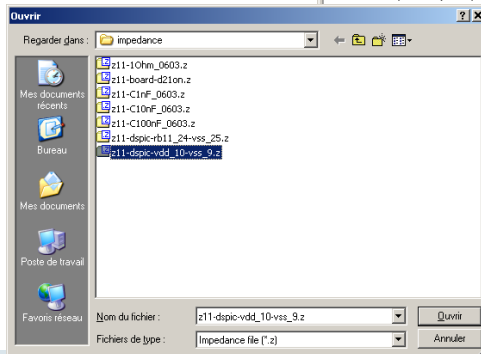
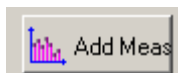
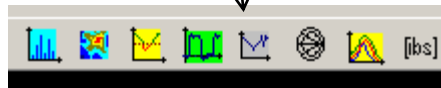


z11-dspic-vdd_10-vss_9.z

Exercise 8. PDN Modelling

- z11-C1nF_0603.z
- 1nF discrete capacitance for DPI

Impedance vs. Freq

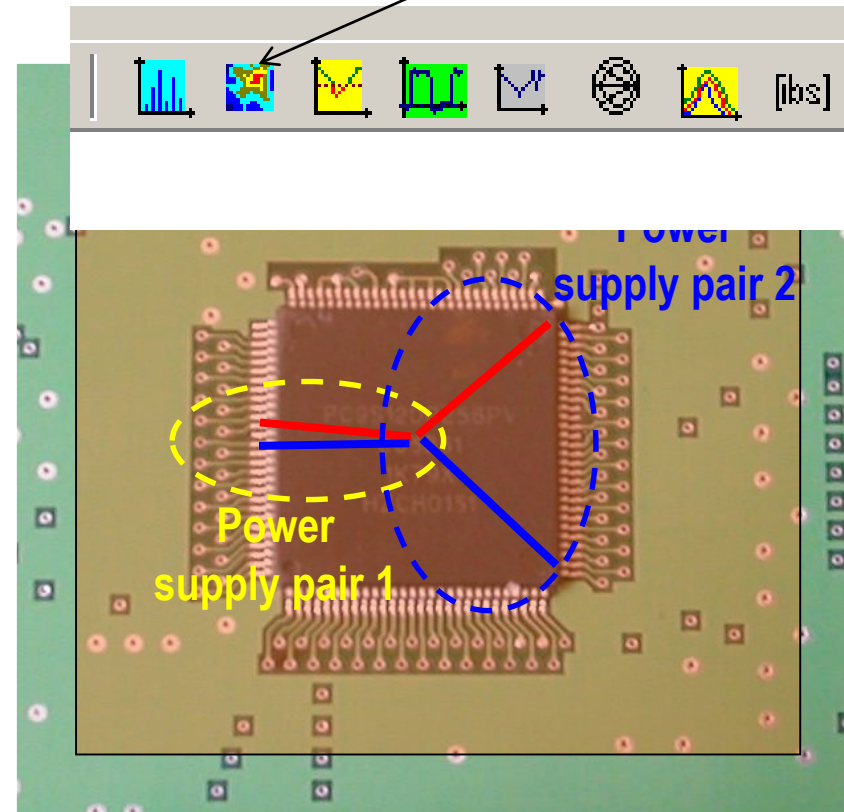


Exercise 9. Radiated Emission Modeling

Near field

- The circuit studied in exercise 5 is mounted in a QFP package.
- Two mounting versions, depending on the power supply pair assignment (pair 1 or pair 2)
- Its magnetic field emission is characterized by near field scan at 100 MHz.
- Compute the magnetic field at 1 mm above the package for both configurations.
- Conclude about the effect of power supply pair placement.

*Ex8-RadEmi_Config1.sch and
Ex8-RadEmi_Config2.sch*



- Package geometry:
 - Width and height = 16 mm
 - Package pitch = 0.5 mm
 - Lead frame height = 0.7 mm

Exercise 9. Radiated Emission Modeling

- Compute the magnetic field at 1 meter above the package for both configurations.
- The following table gives the limit for radiated emission at 1 meter from electronic equipment, defined by CISPR 25.
- Does the circuit complies with CISPR 25 radiated limit at 100 MHz ?

CISPR 25 – Limits for narrowband radiated emission at 1 m from equipment

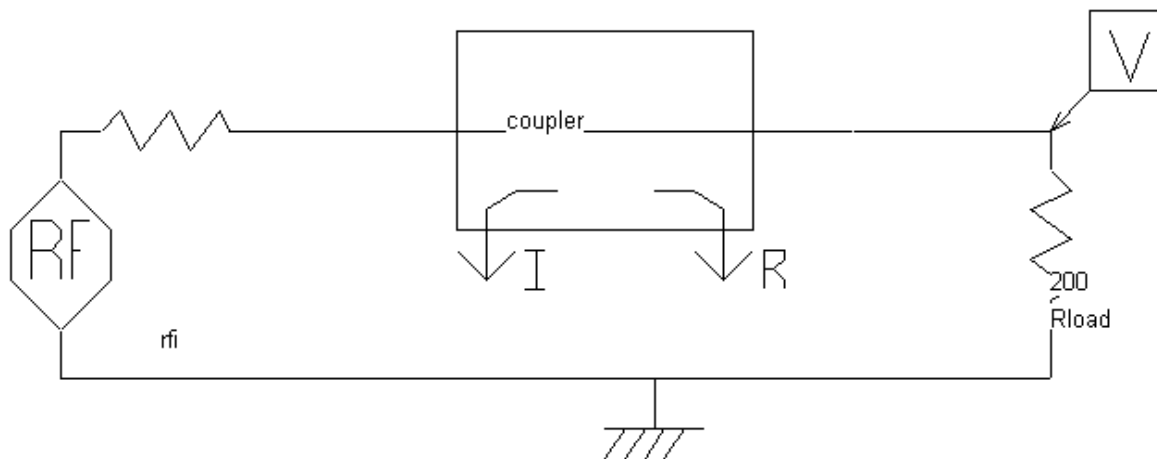
Class	Limit @ 100 MHz
1	42 dB $\mu\text{V}/\text{m}$
2	36 dB $\mu\text{V}/\text{m}$
3	30 dB $\mu\text{V}/\text{m}$
4	24 dB $\mu\text{V}/\text{m}$
5	18 dB $\mu\text{V}/\text{m}$

Synthesis of Exercises 1 to 8

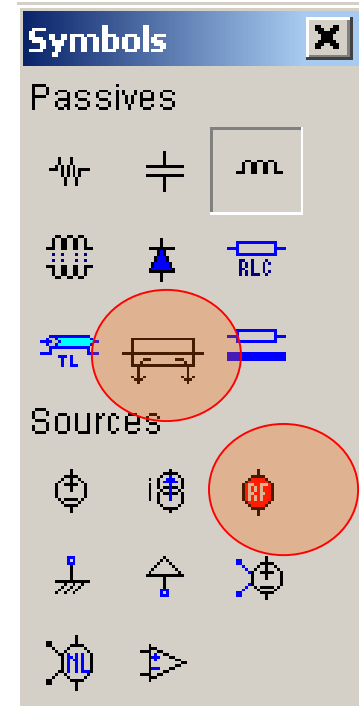
- What did we learn ?

Exercise 10. Estimation of susceptibility level

- A RF generator produces a conducted disturbance which is injected on a $200\ \Omega$ load, though a directional coupler.



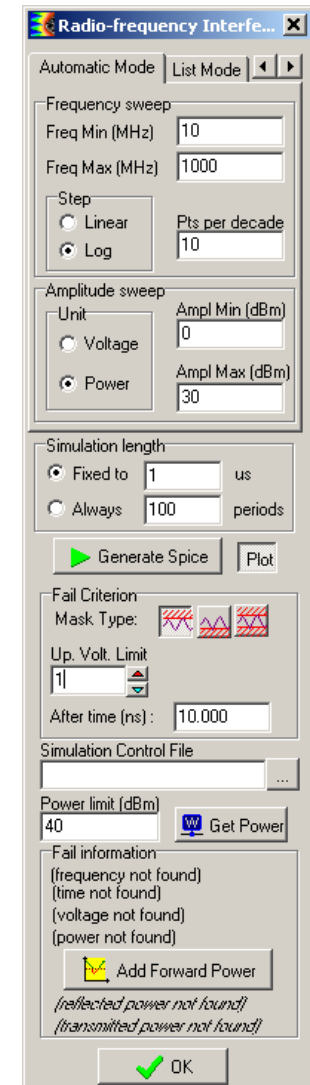
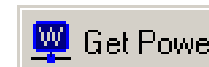
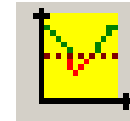
Ex9-RloadSusc.sch



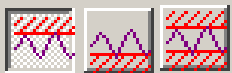
- Estimate the forward power to induce 1 V across the load over the frequency range 10 MHz – 1 GHz.


Exercise 10. Estimation of susceptibility level

- Launch Susceptibility tool
- Configure the RF disturbance and launch SPICE simulation
- Configure the voltage criterion and extract susceptibility threshold
- Display the susceptibility threshold




Exercise 10. Estimation of susceptibility level


Fail Criterion
Mask Type: 


Up. Volt. Limit
1 | 

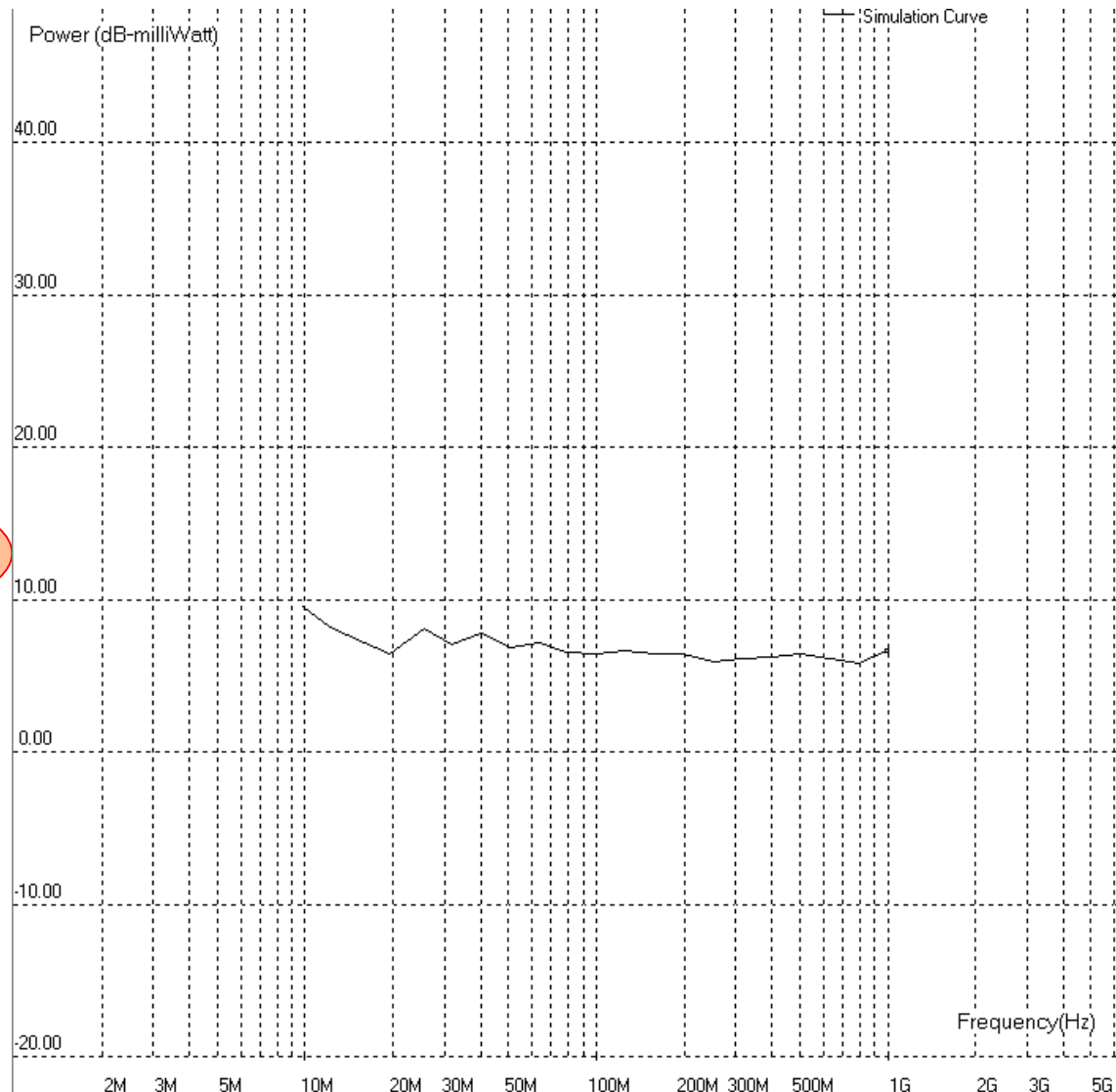
After time (ns) : 10.000

Simulation Control File
D:\Documents and Settings\si ...

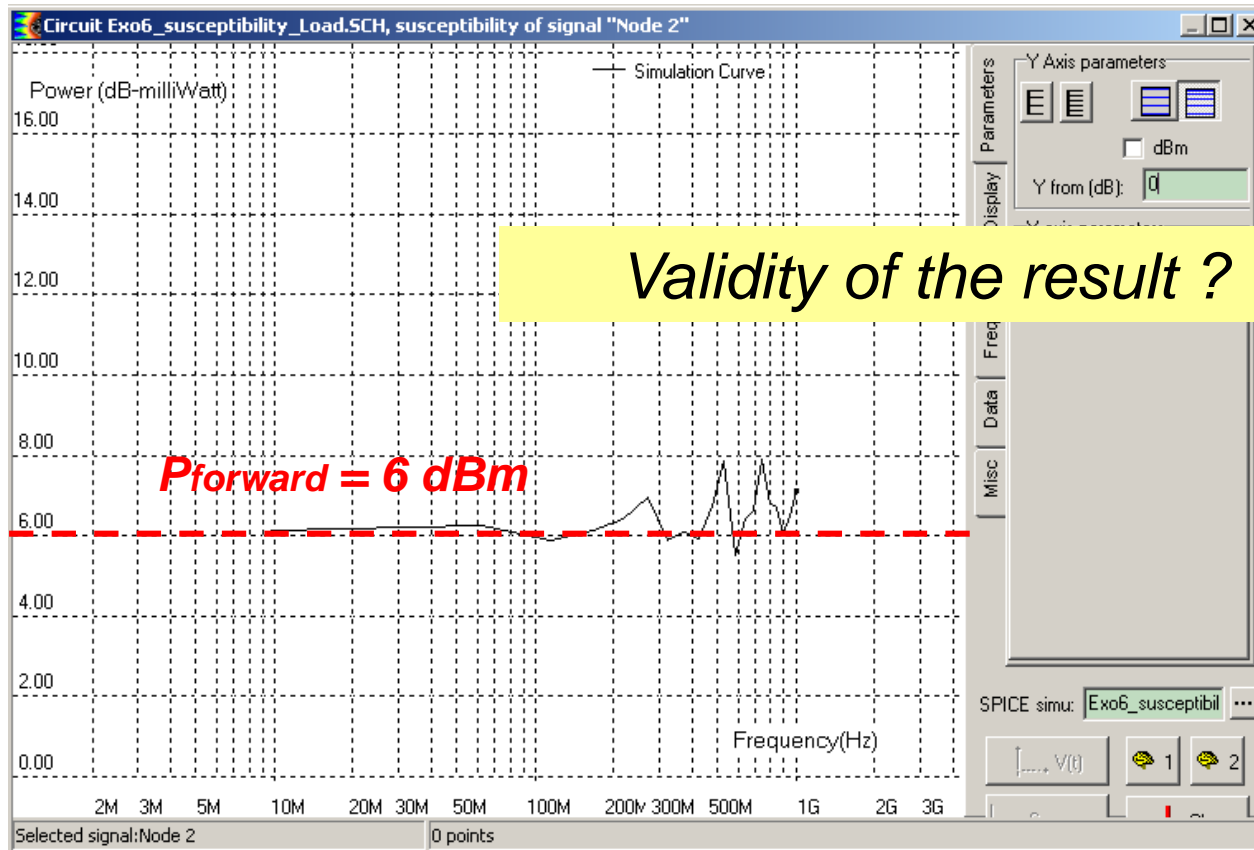
Power limit (dBm)
40 

Fail information
Frfi=1000 MHz
T=61.25 ns
V=1.007 V
Pforward=6.770 dBm (4.754 mW) 
Reflected=1.873 dBm (1.539 mW)
Ptransmitted=5.071 dBm (3.214 mW)

 OK

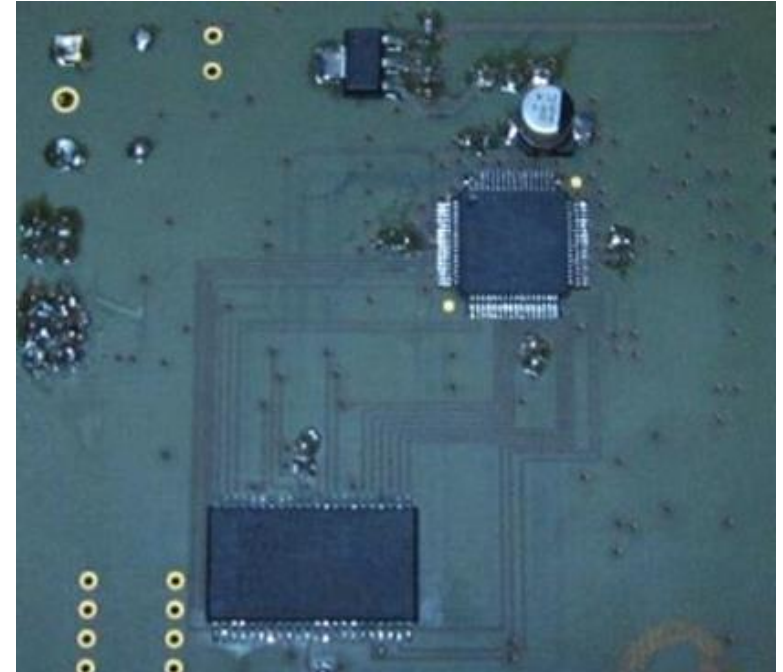


Exercise 10. Estimation of susceptibility level



Exercise 11. Susceptibility of analog input

- A RF disturbance is conducted to an analog input.
- DPI: 1 nF
- PCB: short tracks
- Equivalent model: see IBIS.
- Susceptibility criterion : input noise < 100 mV from 10 MHz to 1 GHz.

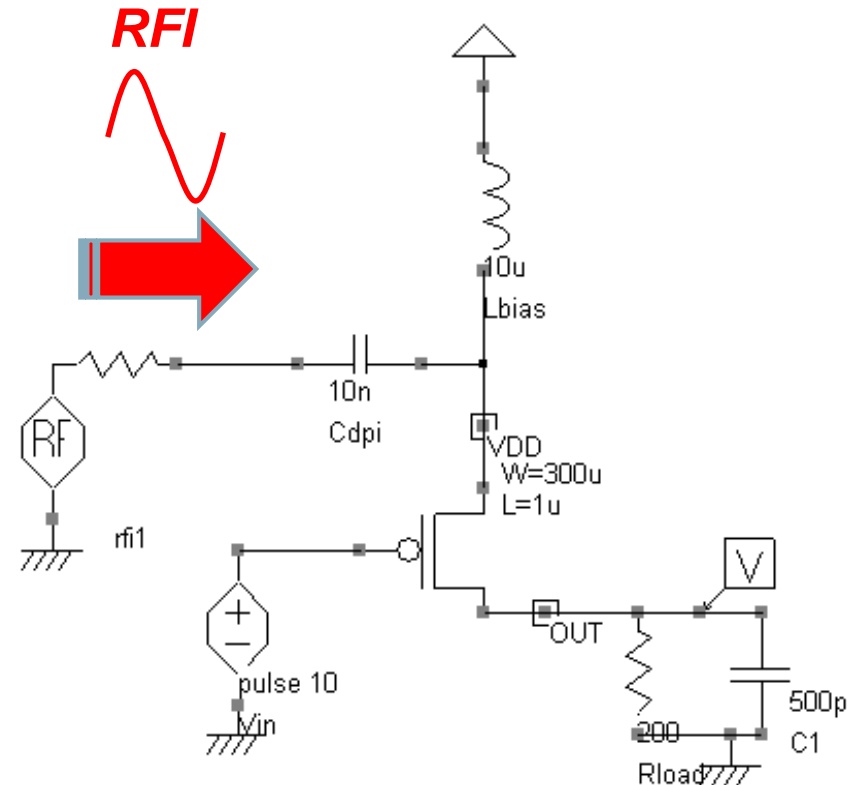


Ex10-ADCInputSusp.sch

Exercise 12. Susceptibility of output buffer

- The following output is loaded by a $200\ \Omega$ load and a $0.5\ \text{nF}$ capacitance.
- The susceptibility of the buffer is tested using DPI standard.
- Harmonic disturbances are injected on power supply

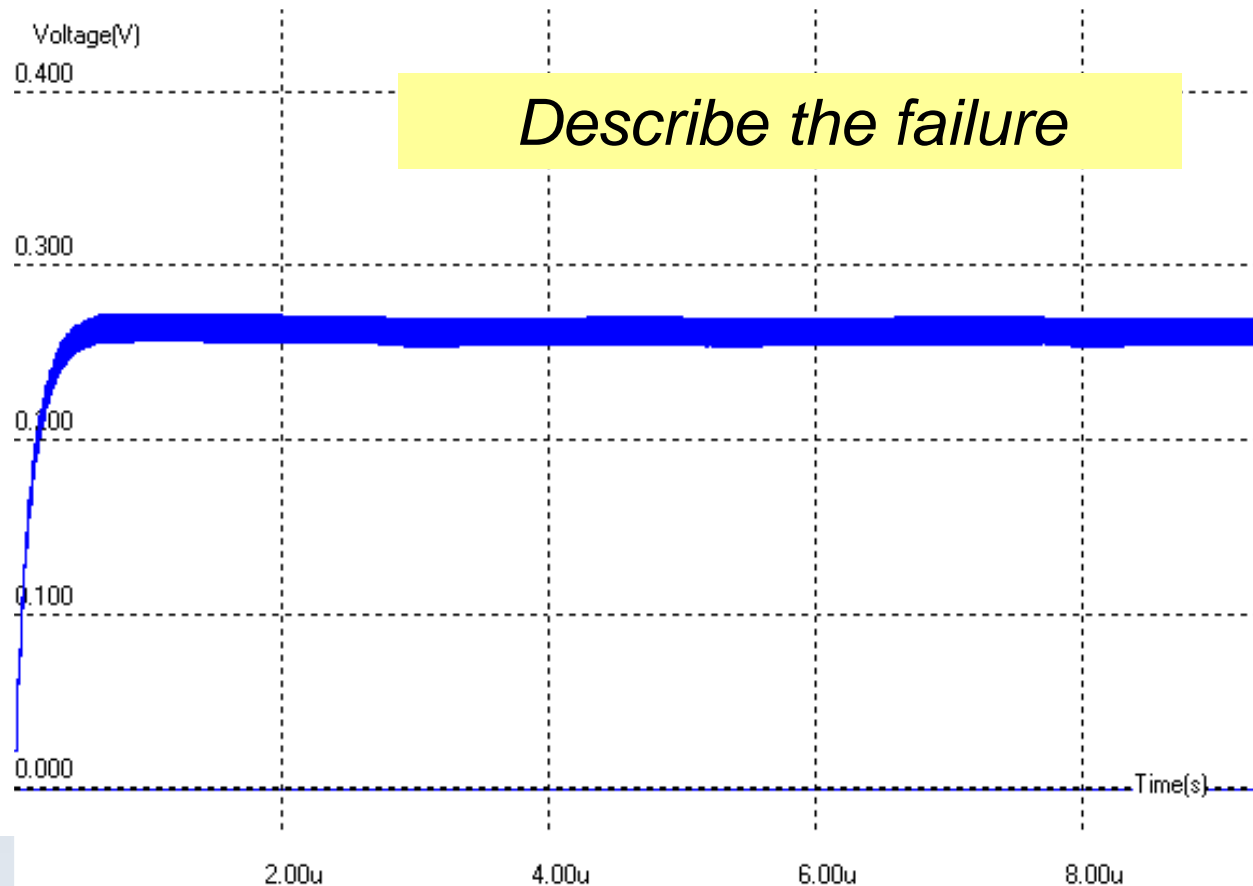
Susceptibility criterion ?



Ex11-HighSideEMI.sch

Exercise 12. Susceptibility of output buffer

- Injection on the Vdd pin
- Add a 100 MHz RFI sinus signal with 2 V amplitude.



Exercise 12. Susceptibility of output buffer

- Failure due to rectification effect.
- Equation of drain current based on MOS Model 1:

$$I_{ds} = \frac{\beta}{2} (V_{GS} - V_T)^2 (1 + \lambda \cdot V_{DS})$$

Square function → Non linear behavior

Synthesis of Exercises 9 to 12

- What did we learn?



Thank you for your attention